

FEATURES

- Wide input range: 66-160VDC
- 250W isolated outputs
- Efficiency up to 91%
- Single output: 24VDC
- Adjustable Vout ($\pm 10\%$)
- Fixed switching frequency, predicted EMI
- Stable @ no-load operation
- Remote On/Off control
- 3KVAC I/O isolation
- Industry standard 1/2nd brick footprint (2.40" × 2.28" × 0.50")
- Extensive self-protection, UVLO, OVP, OTP, OCP and short circuit protection
- Operating temperature range: -40°C to +100°C
- Fully encapsulated, high reliability
- Flexible extra heat-sink mount type
- Compliance with EN 50155 standard



PRODUCT OVERVIEW

The DHB250W110 series use advanced power processing, control and packaging technologies to provide the high performance, flexibility, reliability and cost effectiveness of a mature power converter. Wide range input of 66-160V (110V nominal) that complies with the European EN50155 standard for electronic equipment used on railway rolling stock. Fully encapsulated package technology provides outstanding thermal, vibration & shock performance, is ideal for railway applications where power modules must meet rugged environment requirements.

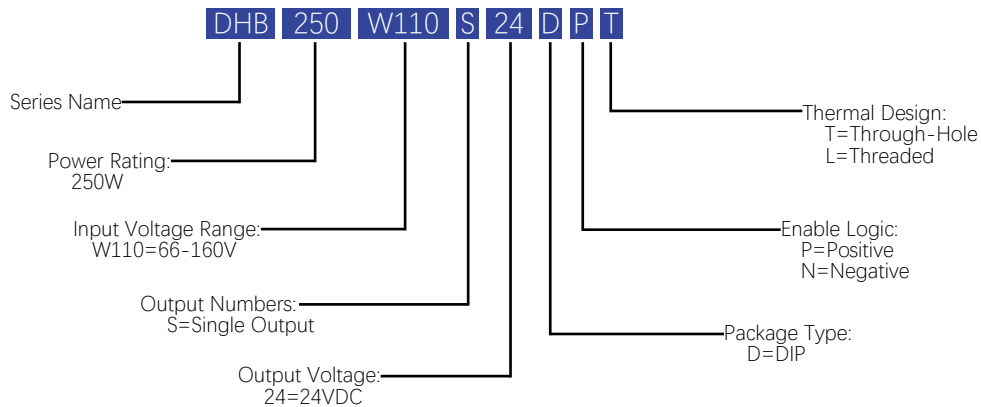
A wealth of self-protection features included input undervoltage lockout, over temperature shutdown, over current protection with "hiccup" autorestart technique, provides indefinite short-circuit protection, along with output OVP. Threaded or through holes are provided to allow easy mount or the addition of a heat sink for extended temperature operation. The operation temperature is -40°C to 100°C, the module delivers full output power @ 100°C case temperature.

The DHB250W110 series are designed to railway standards EN 50155.

Models Selections

| Basic Models | Input Voltage [VDC] | Input Voltage Range [VDC] | Output Voltage [VDC] | Output Current [A] | Efficiency typ. [%] | Capacitive Load Max [μ F] | Package [inch] |
|---------------|---------------------|---------------------------|----------------------|--------------------|---------------------|--------------------------------|----------------|
| DHB250W110S24 | 110 | 66-160 | 24 | 10.4 | 91 | 4000 | 2.40×2.28×0.50 |

Model Numbering



| Absolute Maximum Ratings | | | | | |
|-----------------------------------|----------------------|------|------|--|-------|
| Parameters | Conditions | Min. | Typ. | Max. | Units |
| Input Voltage Continuous | | -0.7 | | 160 | VDC |
| Input Voltage Transient | < 100ms | | | 180 | VDC |
| On/Off Remote Control Voltage | Referred to -Vin | 0 | | 75 | VDC |
| On/Off Remote Control Current | Referred to -Vin | 0 | 0.25 | 1 | mA |
| Operating Baseplate Temperature | | -40 | | 100 | °C |
| Operating Environment Temperature | | -40 | | 85 | °C |
| Storage Temperature Range | | -55 | | 125 | °C |
| Soldering Temperature | Wave Soldering < 10s | | | 260 | °C |
| Safety and EMC Compliance | | | | | |
| Conducted Emission | EN50121-3-2 | | | With external filter | |
| Radiated Emission | EN50121-3-2 | | | With external filter | |
| Conducted Susceptibility | IEC/EN61000-4-6 | | | 10Vrms Criteria A | |
| Radiated Susceptibility | IEC/EN61000-4-3 | | | 10V/m Criteria A | |
| EFT | IEC/EN61000-4-4 | | | ±2KV Criteria A (With external filter) | |
| Surge | IEC/EN61000-4-5 | | | ±2KV Criteria A (With external filter) | |
| ESD | IEC/EN61000-4-2 | | | ±6KV Contact ±8KV Air Criteria A | |

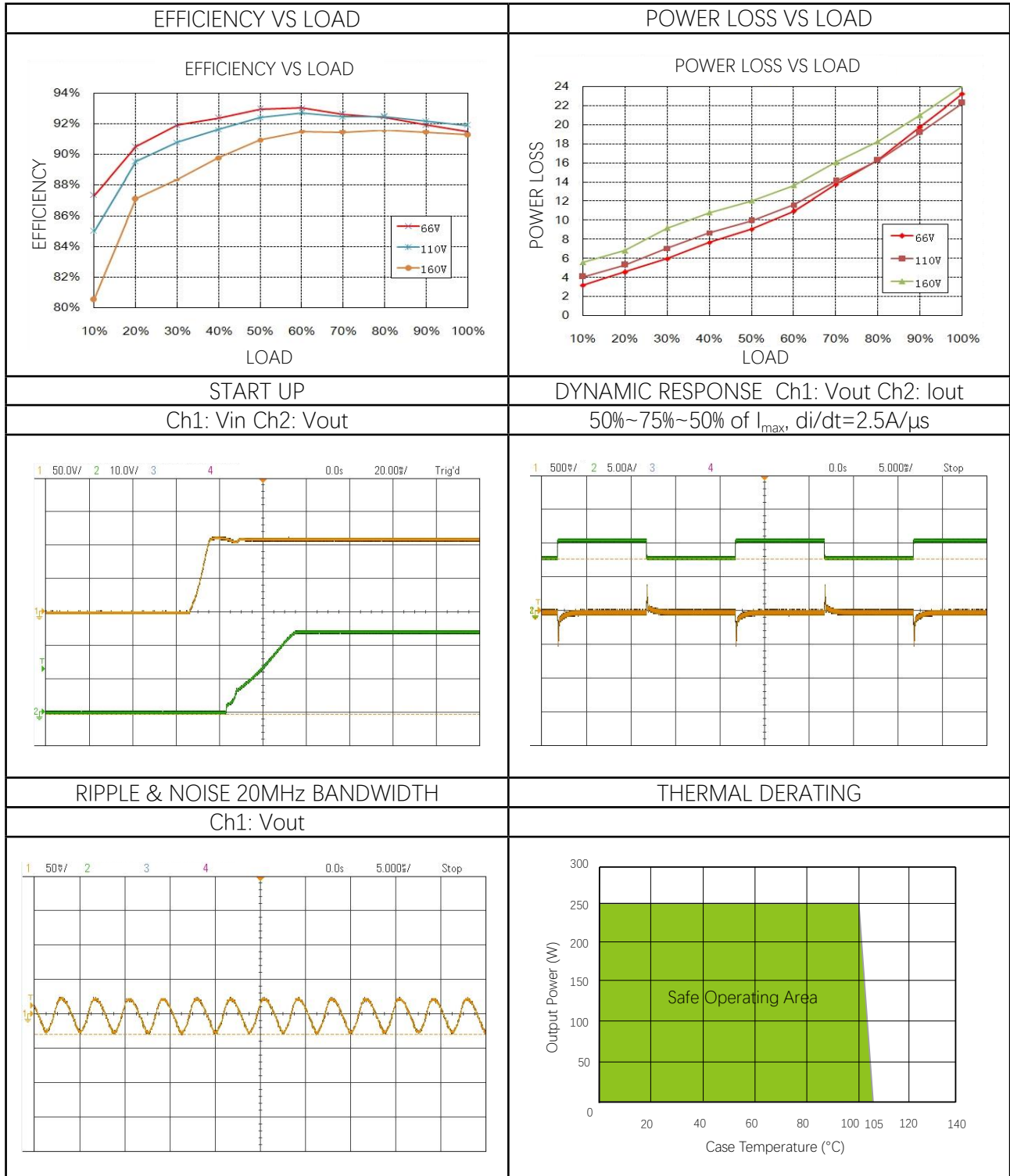
| General Specifications | | | | | | |
|--|--|--------------------------------|------|------|-------|--|
| Parameters | Conditions | Min. | Typ. | Max. | Units | |
| Isolation Voltage | Input to output | | 3000 | | VAC | |
| | Input to case | | 1500 | | VAC | |
| | Output to case | | 1500 | | VAC | |
| Isolation Resistance (Viso=500VDC) | Input to output | 100 | | | MΩ | |
| | Input to case | 100 | | | MΩ | |
| | Output to case | 100 | | | MΩ | |
| Isolation Capacitance | Input to output | | 2200 | | pF | |
| Isolation Safety Rating | Basic insulation | | | | | |
| Switching Frequency | | | 260 | | KHz | |
| Start-up Delay | From start-up threshold recover to 10% Vout | | 30 | 60 | mS | |
| Rise Time | From 10% Vout to 90% Vout capacitive load | | 26 | 50 | mS | |
| Remote On/Off Control | Positive Logic, ON state | Open or $3 \leq V_r \leq 75$ | | | VDC | |
| | Positive Logic, OFF state | Short or $0 \leq V_r \leq 1.2$ | | | VDC | |
| | Negative Logic, ON state | Short or $0 \leq V_r \leq 1.2$ | | | VDC | |
| | Negative Logic, OFF state | Open or $3 \leq V_r \leq 75$ | | | VDC | |
| Remote Control Current | | 0 | 0.25 | 1 | mA | |
| Vibration | IEC61373:1999 Category I, Class B, Body mounted | | | | | |
| Shock | IEC61373:1999 Category I, Class B, Body mounted | | | | | |
| Input Specifications | | | | | | |
| Parameters | Conditions | Min. | Typ. | Max. | Units | |
| Operating Voltage Range | | 66 | 110 | 160 | VDC | |
| Start-up Threshold | | 62 | 64 | 66 | VDC | |
| Under Voltage Shutdown | | 58 | 60 | 62 | VDC | |
| Input Over Voltage Protection | | 170 | 178 | 185 | VDC | |
| Input Over Voltage Recover | | 162 | 170 | 178 | VDC | |
| Input Current @ No Load | | | 30 | 100 | mA | |
| Input Current @ Min. Line | Vin=min.line, Iout=full load | | | 4.5 | A | |
| Input Current @ Shutdown Mode | | | 15 | 50 | mA | |
| Reflect Ripple Current (Peak-Peak) | Measured at input pin with 10μH inductor and 100μF capacitance | | 180 | 240 | mA | |
| Recommended Input Fuse | | | 9 | | A | |
| Recommended External Input Capacitance | 1μF CBB and 100μF E-cap used in combination | | 100 | | μF | |

Performance Data (24 Vout)

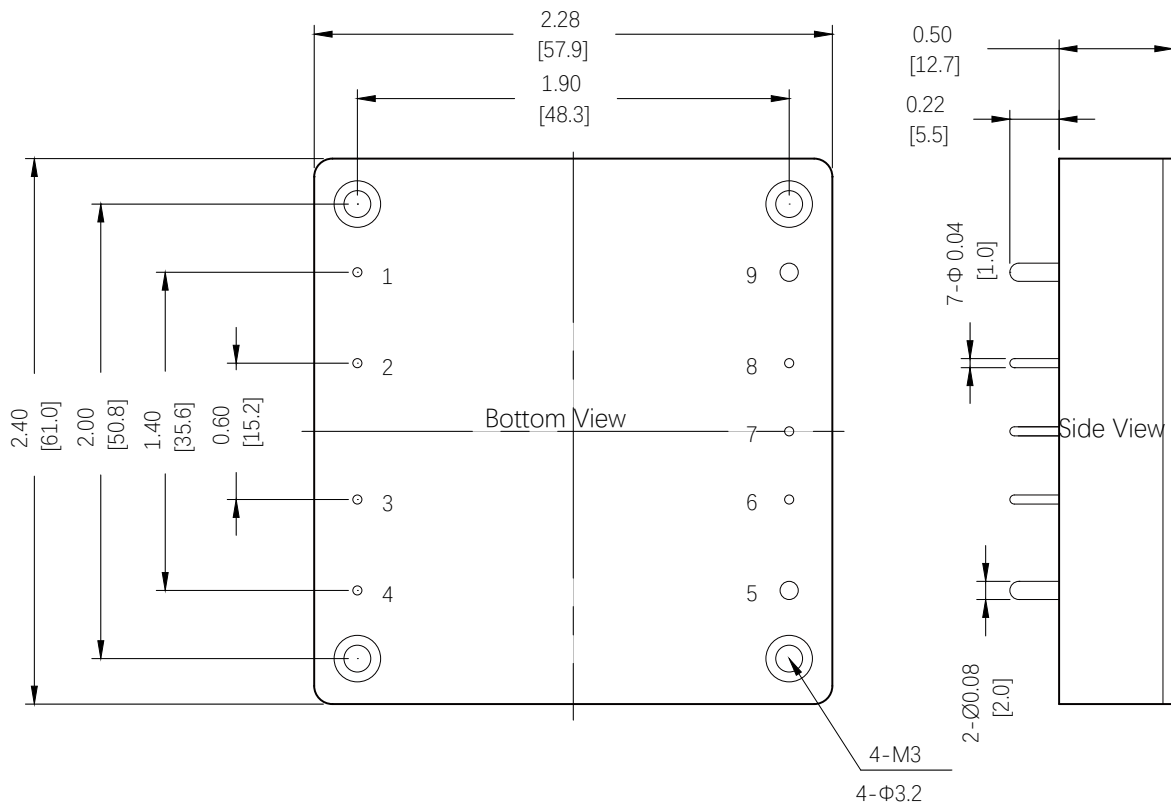
| Output Specifications | | | | | |
|---|-----------------------------|-------|-------|-------|---------------|
| Parameters | Conditions | Min. | Typ. | Max. | Units |
| Output Power | | | | 250 | W |
| Output Voltage Setpoint | Nom.line, 50% Load | 23.64 | 24.00 | 24.36 | V |
| Vout Accuracy | | -1.5 | | +1.5 | % of Vout |
| Adjustable Range | Trim up/ Trim down | -10 | | +10 | % of Vout |
| Line Regulation | | -0.2 | | +0.2 | % |
| Load Regulation | | -0.5 | | +0.5 | % |
| Temperature Coefficient | | -0.02 | | +0.02 | % of Vout /°C |
| Total Regulation | | -3 | | +3 | % |
| Thermal Shutdown | | 100 | 105 | 115 | °C |
| Over Voltage Protection | Hiccup | 115 | | 140 | % of Vout |
| Over Current Protection | Hiccup | 110 | | 160 | % of Iout |
| Short Circuit Protection | Hiccup | | | | |
| Remote Sense Voltage | | | | 10 | % |
| Ripple & Noise Max. ^① | | | | 240 | mV pk-pk |
| Dynamic Load Peak Deviation ^② | | -5 | | +5 | % of Vout |
| Dynamic Load Response | | | | 500 | μS |
| Capacitive Load | | 100 | | 4000 | μF |
| Minimum Load | No minimum load requirement | | | | |
| Notes | | | | | |
| ① Ripple & noise is tested with certain filter parameters, please see output ripple & noise in technical notes on page 8 for more details. | | | | | |
| ② The load is set from 75%-100%-75% of I _{max} , di/dt=2.5A/μS, Cout=100μF, please refer to dynamic waveforms in performance data on page 5 for details. | | | | | |

All specifications are tested at 25 °C ambient temperature, nominal input voltage, rated output current conditions unless otherwise specified.

Performance Data (24 Vout)



Mechanical Specifications



PIN:

PIN1, PIN2, PIN3, PIN4 PIN6, PIN7, PIN8: Φ0.04inch

Force: Applied force not exceed 4.9N

PIN5, PIN9 : Φ0.08inch

Force: Applied force not exceed 9.8N

Material: Copper alloy

Finish: Gold 3 ~ 5µm(min.) over nickel 50µm(Min.)

TOLERANCE:

X.XX=±0.02[0.5]

X.XXX= ±0.010[0.25]

Dimensions are in inches [mm]

Weight: ~115g.

| PIN CONNECTIONS | |
|-----------------|----------|
| Single Output | |
| Pin | Function |
| 1 | -Vin |
| 2 | NC |
| 3 | RC |
| 4 | +Vin |
| 5 | +Vout |
| 6 | +Sense |
| 7 | TRIM |
| 8 | -Sense |
| 9 | -Vout |

Technical Notes

INPUT FUSING

Certain applications may require fuse at the inputs of power conversion components. Fuses should also be used when there is possibility of sustained input voltage reversal which is not current limited. The DHB250W110 modules are not internally fused. We strongly recommend a slow-blown fuse to be used in the ungrounded input supply line.

For safety agency approvals, the installer must install the converter in compliance with the end user safety standard.

TYPICAL APPLICATION CONNECTION

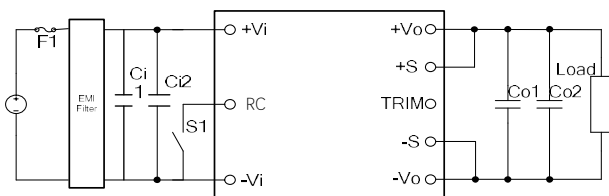


Figure 1: Typical Application Connection

Figure 1 shows the typical use of the module connection. In order to prevent the input line from causing the input oscillation, it is recommended to add the input capacitor close to the input pin of the module. Similarly, the output capacitor is added to the output of the module. Specific recommended parameters: input capacitance $C_{i1}=100\mu\text{F}$ electrolytic capacitor, $C_{i2} = 1\mu\text{F}$ CBB capacitor. Output Capacitance $C_{o1}=10\mu\text{F}$ tantalum capacitor, C_{o2} ESR $<0.1\Omega$.

REFLECTED RIPPLE CURRENT

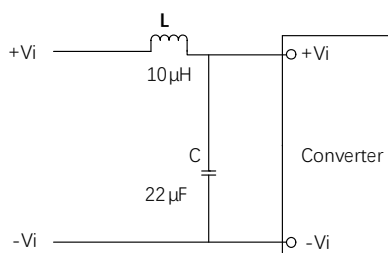


Figure 2: Reflected Ripple Current

Add LC filter at the front of the power module to reduce the interference of reflected ripple current on the DC bus, recommended value of L and C with appropriate current and voltage rating as below: $L=10\mu\text{H}$; $C=100\mu\text{F}$.

REMOTE CONTROL FUNCTION

Module Power Remote Control or called ON/OFF pin is for the user to enable or disable the output. Control use high and low level control, there are two general control logic, positive logic or negative logic control. Recommend to use optocoupler to control ON/OFF Pin as below.

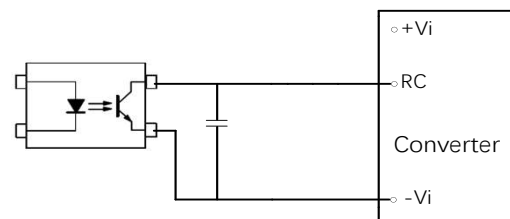


Figure 3: Remote Control

Remote Control Pin can be connected in parallel for multiple converters which with the same Remote Control characters. However, when several converters share the same remote control circuit, the total sink and source current must be taken into consideration, and make sure that the optocoupler has enough drive capability.

To reduce external PCB trace interference, it is recommended to add high frequency bypass capacitor between RC pin and -Vi, recommended capacitor value is 100-1000pF.

REMOTE COMPENSATION FUNCTION

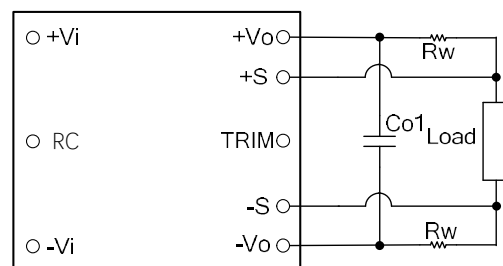


Figure 4: Remote Compensation

Technical Notes

The remote compensation function compensates for the voltage drop across the output line. Module compensation function can't exceed 10%, that is:

$$[(+VO) - (-VO)] - [(+S) - (-S)] \leq 10\%V_{onom}$$

If the remote compensation function is not used, the +Sense and +Vout pin, -Sense and -Vout pin need to be shorted directly close to the output.

OUTPUT RIPPLE & NOISE

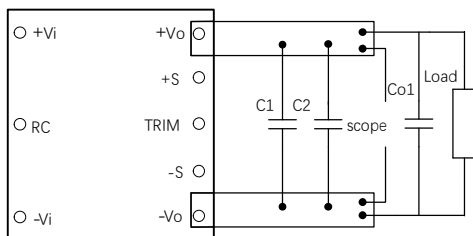


Figure 5- Output Ripple

These DHB250W110 modules' output ripple and noise is measured at the rated input voltage and output current, along with 10uF MLCC capacitor and 0.1uF MLCC used in parallel with appropriate voltage ratings and placed as C1&C2 shown in the figure above. The scope's bandwidth is set to 20MHz.

External output capacitors are required to reduce the ripple & noise. The output capacitors should be low ESR and appropriate frequency response with appropriate voltage ratings, and must be located as close to the converters as possible, also PCB layout must be taken into consideration.

INPUT UNDERVOLTAGE SHUTDOWN AND START-UP THRESHOLD

Under normal start-up conditions, module will not begin to regulate until the ramping-up input voltage exceeds the Start-Up Threshold Voltage. Once operating, module will not turn off until the input voltage drops below the Undervoltage Shutdown limit. Subsequent re-start will not occur until the input is brought back up to the Start-Up Threshold. This built in hysteresis prevents any

unstable on/off situations from occurring at a single input voltage.

CURRENT LIMITING

The maximum current limit remains constant as the output voltage drops. However, once the impedance of the short across the output is small enough to make the output voltage drop below the specified Output DC Current Limit Shutdown Voltage, the converter turns off.

The converter then enters a "hiccup mode" where it repeatedly turns on and off until the short circuit condition is removed. This prevents excessive heating of the converter or the load board.

SHORT CIRCUIT CONDITION

When the converter is in current-limit mode, the output voltage will drop as the output current demand increases and then the converter will be shut down. If the short-circuit condition persists, another shutdown cycle will be initiated. This on/off cycling is referred to as "hiccup" mode. The hiccup cycling reduces the average output current, thereby preventing internal temperatures from rising to excessive levels. The module is capable of enduring an indefinite short circuit output condition.

OUTPUT OVERVOLTAGE PROTECTION

The output voltages are monitored for an overvoltage condition via magnetic feedback. The signal is coupled to the primary side and if the output voltage rises to a level which could be damaging to the load, the sensing circuitry will power down the PWM controller causing the output voltages to decrease. Following a timeout period the PWM will restart, causing the output voltages to ramp to their appropriate values. If the fault condition persists, and the output voltages again climb to excessive levels, the overvoltage circuitry will initiate another shutdown cycle. This on/off cycling is referred to as "hiccup" mode.

Technical Notes

THERMAL SHUTDOWN

These DHB250W110 converters are equipped with thermal shutdown circuitry. If environmental conditions cause the internal temperature of the DC-DC converter to rise above the designed operating temperature, a precision temperature sensor will power down the unit. When the internal temperature decreases below the threshold of the temperature sensor, the unit will auto restart.

TRIMMING OUTPUT VOLTAGE

DHB250W110 converters have a trim capability that allows users to adjust the output voltages. Output voltage can be trimmed up or down by a trim pin by connecting a single fixed resistor between Trim Pin and Vout+ or Vout-, the output voltage can be increased or decreased depending on its connection. The maximum output voltage adjustment range is -10% to +10%. If the trim function is not used, keep TRIM pin floating.

Trim up:

Add a fixed resistor between in TRIM and -Vo, you can achieve the output voltage up. Do not trim the converter above maximum trim limit (typically 10%) or the output over voltage protection circuit may be activated.

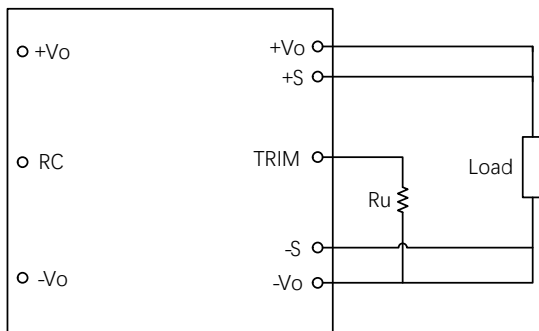


Figure 6· Trim Up Connection

Please follow up the Trim Up formula to calculate the resistor value according to the desired output voltage.

$$R_u = \frac{2.48 \times [R1 \times R2 + 5.6 \times (R1 + R2)] - 5.6 \times V_{oset} \times (1 + \Delta) \times R2}{V_{oset} \times (1 + \Delta) \times R2 - 2.48 \times (R1 + R2)} \quad (\text{k}\Omega)$$

Where:

R1=20KΩ,

R2=2.32KΩ,

"Voset" is the output voltage when TRIM is floating, "Δ" is the change of output voltage, such as: 15V output is raised to 16.5V,

$$\Delta = (16.5 - 15) / 15 * 100\% = 10\%$$

Trim down:

Add a fixed resistor between in TRIM and +Vo, you can achieve the output voltage down. Do not exceed maximum rated output load.

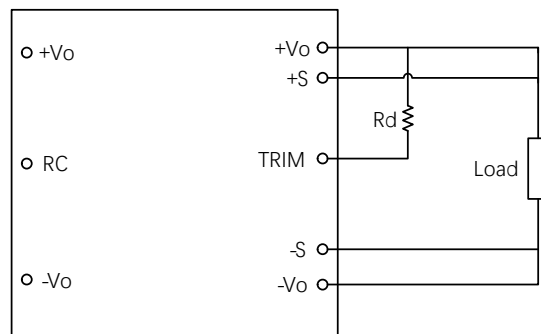


Figure 7· Trim Down Connection

Please follow up the Trim Down formula to calculate the resistor value according to the desired output voltage.

$$R_d = \frac{2.48 \times 5.6 \times R1 - R2 \times (R1 + 5.6) \times [V_{oset} \times (1 - \Delta) - 2.48]}{V_{oset} \times (1 - \Delta) \times R2 - 2.48 \times (R1 + R2)} \quad (\text{k}\Omega)$$

Where:

R1=20KΩ,

R2=2.32KΩ,

"Voset" is the output voltage when TRIM is floating, "Δ" is the change of output voltage, such as: 15V output is reduced to 13.5V,

$$\Delta = (15 - 13.5) / 15 * 100\% = 10\%$$

Technical Notes



This product is subject to the following operating requirements and the Life and Safety Critical Application Sales Policy:

Refer to: <http://www.densitypower.com>

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