

FEATURES

- Wide input range: 43-160VDC
- 100W isolated outputs
- Efficiency up to 90%
- Fixed outputs from 5, 12, 15 and 24VDC
- Adjustable Vout ($\pm 10\%$)
- Fixed switching frequency, predicted EMI
- Stable @ no-load operation
- Remote On/Off control
- 1500VAC I/O isolation
- Industry standard 1/4th brick footprint (2.28"×1.45"×0.50")
- Extensive self-protection, UVLO, OVP,OTP,OCP and short protection
- Operating temperature range: -40°C to +100°C
- Fully encapsulated, high reliability
- Flexible extra heat-sink mount type
- Compliance with EN50155 standard



PRODUCT OVERVIEW

The DQB100W110 series use advanced power processing, control and packaging technologies to provide the high performance, flexibility, reliability and cost effectiveness of a mature power converter. Wide input range of 43-160V(110V nominal) that complies with the European EN50155 standard for electronic equipment used on railway rolling stock. Fully encapsulated package technology provides outstanding thermal, vibration & shock performance, is ideal for railway applications where power modules must meet rugged environment requirements.

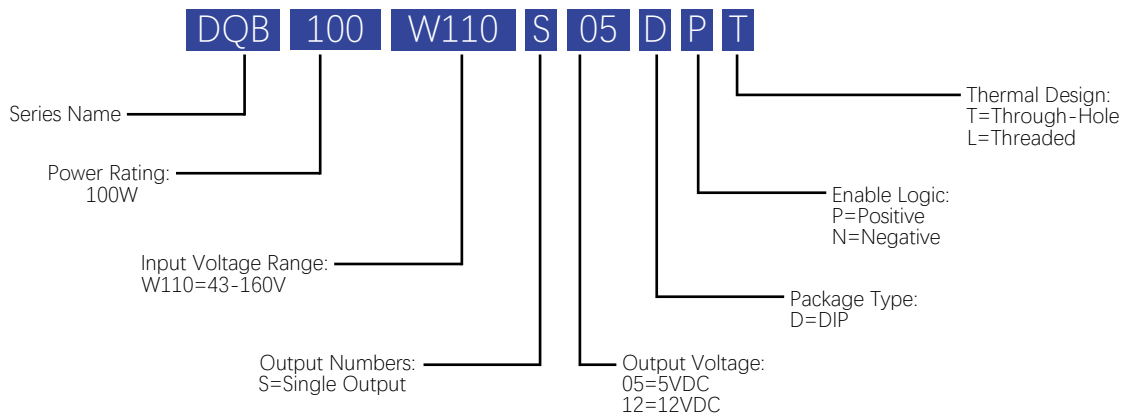
A wealth of self-protection features included input undervoltage lockout, over temperature shutdown, over current protection with "hiccup" autorestart technique, provides indefinite short-circuit protection, along with output OVP. Threaded or through holes are provided to allow easy mount or the addition of a heat sink for extended temperature operation. The operating temperature is -40°C to 100°C, the module delivers full output power @ 100°C baseplate temperature.

The DQB100W110 series are designed to railway standards EN 50155.

Models Selections

Basic Models	Input Voltage [VDC]	Input Voltage Range [VDC]	Output Voltage [VDC]	Output Current [A]	Efficiency typ. [%]	Capacitive Load Max [μ F]	Package [inch]
DQB100W110S05D	110	43-160	5	20	85	6800	2.28"×1.45"×0.50"
DQB100W110S12D	110	43-160	12	8.34	89	2700	
DQB100W110S15D	110	43-160	15	6.67	90	2200	
DQB100W110S24D	110	43-160	24	4.17	90	1500	

Model Numbering



Absolute Maximum Ratings						
Parameters	Conditions	Min.	Typ.	Max.	Units	
Input Voltage Continuous		-0.7		160	VDC	
Input Voltage Transient	< 100ms			180	VDC	
On/Off Remote Control Voltage	Referred to -Vin	0		75	VDC	
On/Off Remote Control Current		0	0.25	1	mA	
Operating Baseplate Temperature		-40		100	°C	
Operating Environment Temperature		-40		85	°C	
Storage Temperature Range		-55		125	°C	
Soldering Temperature	Wave Soldering < 10s			260	°C	
Safety and EMC Compliance						
Conducted Emission	EN50121-3-2	With external filter				
Radiated Emission	EN50121-3-2	With external filter				
Conducted Susceptibility	IEC/EN61000-4-6	10Vrms Criteria A				
Radiated Susceptibility	IEC/EN61000-4-3	10V/m Criteria A				
EFT	IEC/EN61000-4-4	±2KV Criteria A (With external filter)				
Surge	IEC/EN61000-4-5	±2KV Criteria A (With external filter)				
ESD	IEC/EN61000-4-2	±6KV Contact ±8KV Air Criteria A				

General Specifications						
Parameters	Conditions	Min.	Typ.	Max.	Units	
Isolation Voltage	Input to output	1500			VAC	
	Input to case	1500			VAC	
	Output to case	1500			VDC	
Isolation Resistance (Viso=500VDC)	Input to output	100			MΩ	
	Input to case	100			MΩ	
	Output to case	100			MΩ	
Isolation Capacitance	Input to output		1500		pF	
Isolation Safety Rating	Basic insulation					
Switching Frequency			250		KHz	
Start-up Delay	From start-up threshold recover to 10% Vout			80	mS	
Rise Time	From 10% Vout to 90% Vout capacitive load			40	mS	
Remote On/Off Control						
"P" suffix						
Positive Logic, ON state		Open or $3.0 \leq V_r \leq 75$			VDC	
Positive Logic, OFF state		Short or $0 \leq V_r \leq 1.2$			VDC	
"N" suffix						
Negative Logic, ON state		Short or $0 \leq V_r \leq 1.2$			VDC	
Negative Logic, OFF state		Open or $3.0 \leq V_r \leq 75$			VDC	
Remote Control Current		0	0.25	1.5	mA	
Vibration	IEC61373:1999 Category I, Class B, Body mounted					
Shock	IEC61373:1999 Category I, Class B, Body mounted					
Input Specifications						
Parameters	Conditions	Min.	Typ.	Max.	Units	
Operating Voltage Range		43	110	160	VDC	
Start-up Threshold		39		43	VDC	
Under Voltage Shutdown		36		41	VDC	
Input Current @ No Load			30	50	mA	
Input Current @ Min. Line	Vin=min.line, Iout=full load			3	A	
Input Current @ Shutdown Mode			6	10	mA	
Input Reflected Ripple Current (Peak-Peak)	Measured with 10μH inductor and 22μF capacitance		50	100	mA	
Recommended Input Fuse			5		A	
Recommended External Input Capacitance	1μF CBB and 22μF E-cap used in combination	47	100		μF	

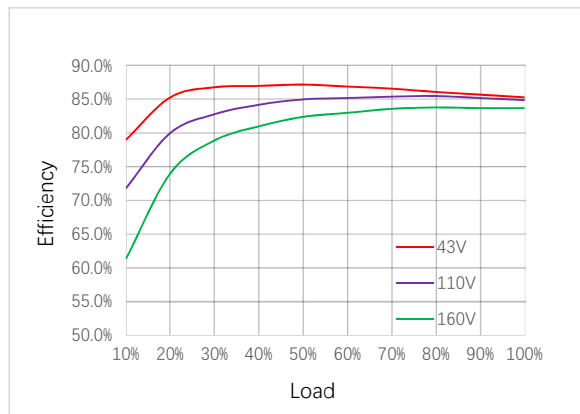
Output Specifications					
Parameters	Conditions	Min.	Typ.	Max.	Units
Output Power				100	W
Vout Accuracy	50% Load, Vin nom	-1.5		+1.5	% of V
Adjustable Range	Trim up/ Trim down	-10		+10	% of Vout
Line Regulation	Vin=43-160VDC, half load	-0.2		+0.2	%
Load Regulation	Vin=110VDC, load=0%-100% of full load	-0.5		+0.5	%
Temperature Coefficient		-0.02		+0.02	% of Vout /°C
Total Regulation		-2		+2	%
Thermal Shutdown		105	110	115	°C
Thermal Shutdown Recover		85	90	95	°C
Over Voltage Protection	Hiccup	115		140	%Vout
Over Current Protection	Hiccup	110		160	%Iout
Short Circuit Protection	Hiccup				
Remote Sense Voltage				10	%
Minimum Load	No minimum load required				
Output Specifications					
Parameters	Modules				
	S05	S12	S15	S24	
Output Voltage Normal(VDC)	5.0	12	15	24	
Ripple & Noise Max. (mV pk-pk) ^①	100	120	150	240	
Dynamic Load Peak Deviation (%Vout) ^②	±5	±5	±5	±5	
Dynamic Load Response (µS)	500	500	500	500	
Capacitive Load (µF)	Min	220	100	100	100
	Max	6800	2700	2200	1500
Notes					
① Ripple & noise is tested with certain filter parameters, please see output ripple & noise in technical notes on page 12 for more details.					
② The load is set from 75%-100%-75% of I _{max} , di/dt=1A/µS, please refer to dynamic waveforms in performance data on page 8 for details.					

All specifications are tested at 25 °C ambient temperature, nominal input voltage, rated output current conditions unless otherwise specified.

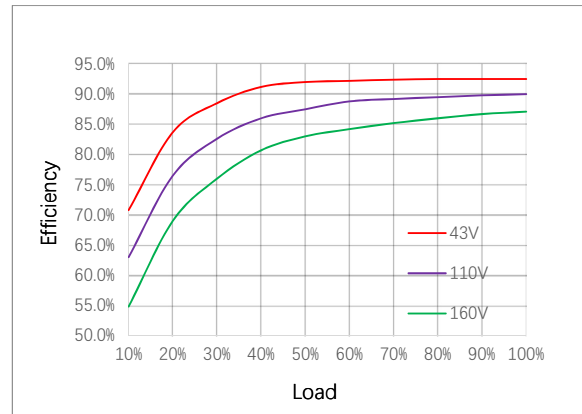
Performance Data

EFFICIENCY VS LOAD

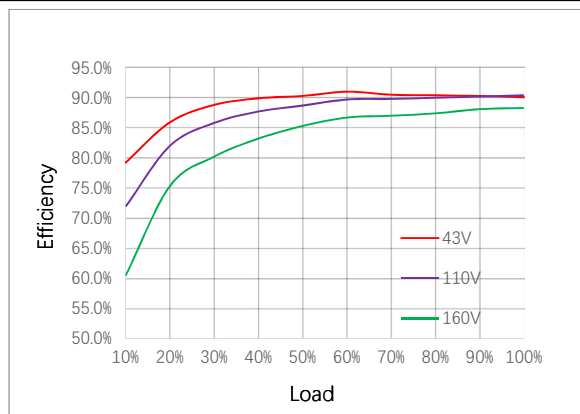
DQB100W110S05



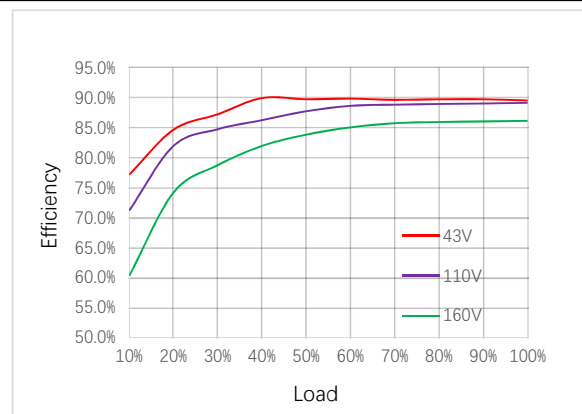
DQB100W110S12



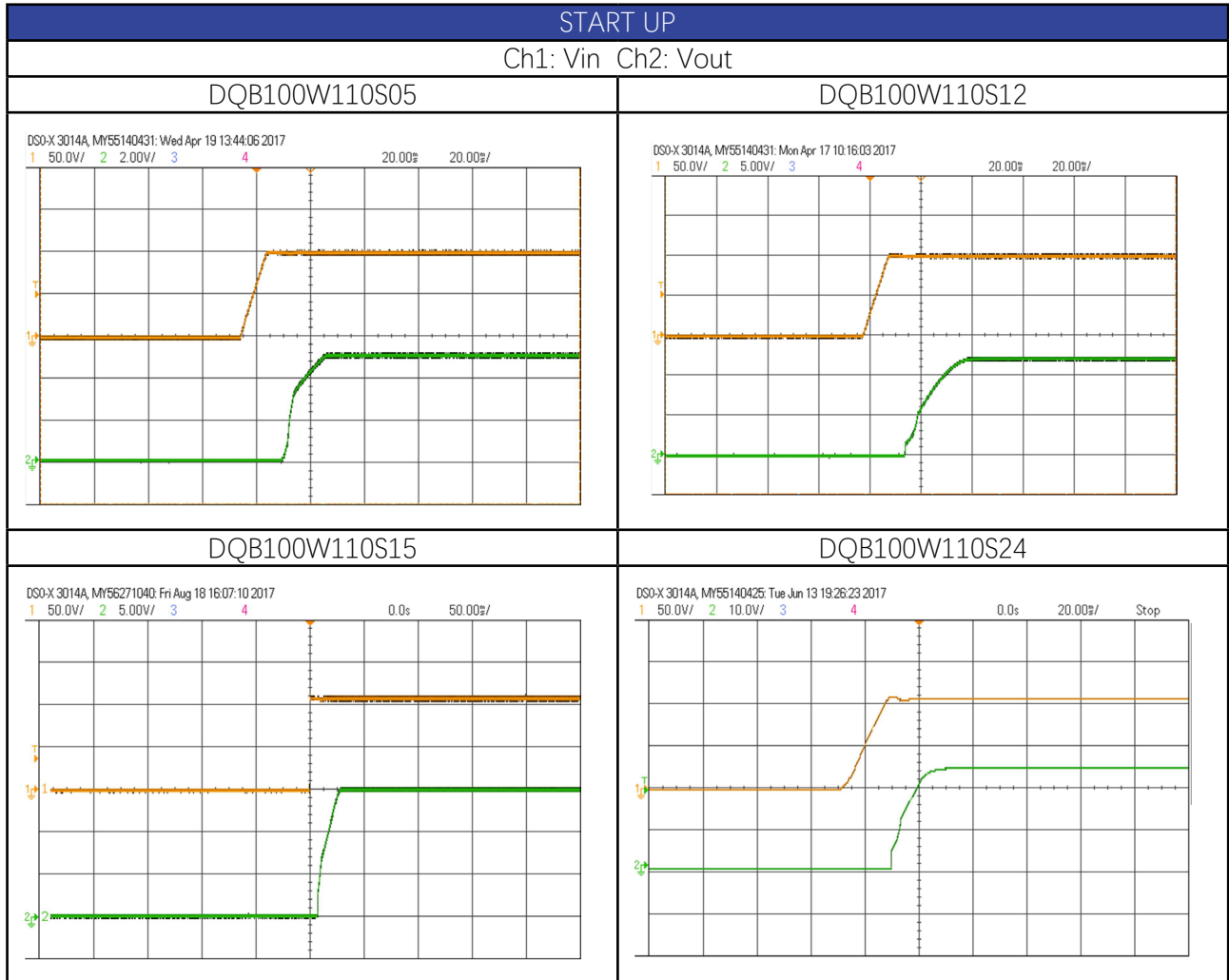
DQB100W110S15



DQB100W110S24



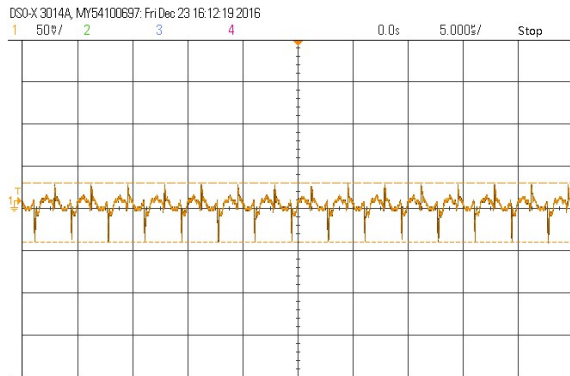
Performance Data



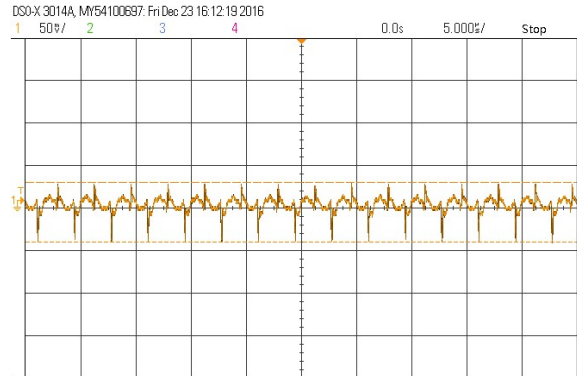
Performance Data

RIPPLE & NOISE 20MHz Bandwidth

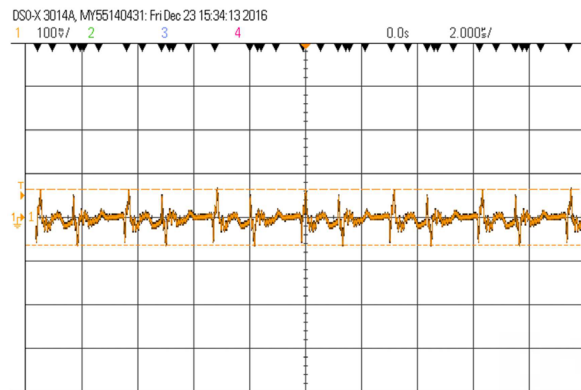
DQB100W110S05



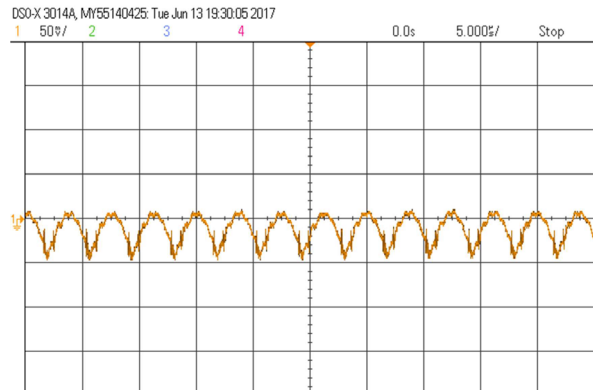
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DQB100W110S15



DQB100W110S24

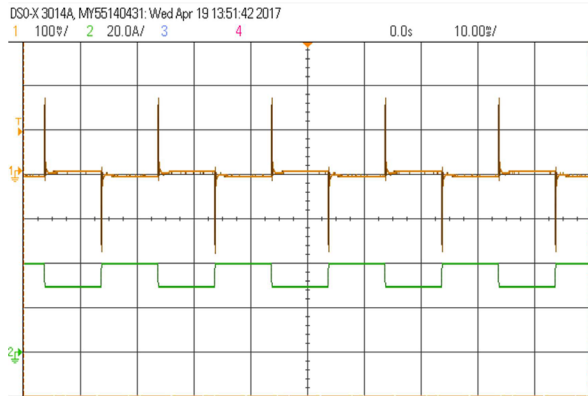


Performance Data

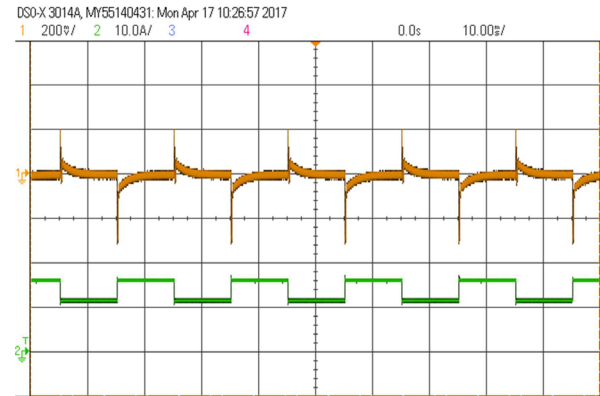
DYNAMIC RESPONSE (75%~100%~75% of I_{max} , $di/dt=1A/\mu s$)

Ch1: V_{out} Ch2: I_{out}

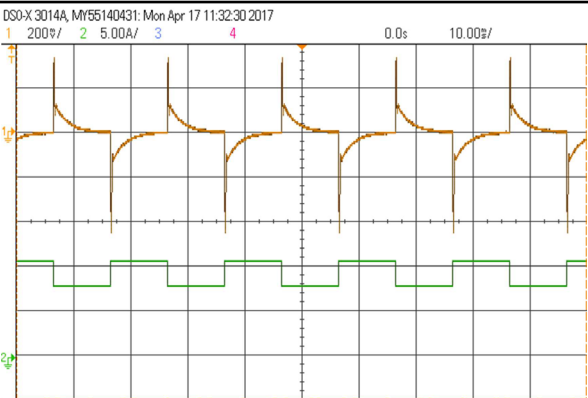
DQB100W110S05 $C_{out}=470\mu F$



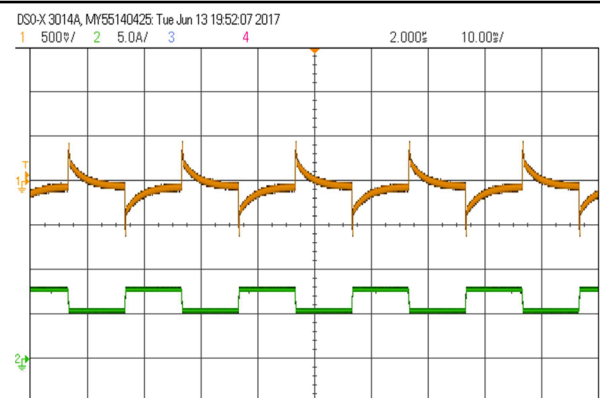
DQB100W110S12 $C_{out}=100\mu F$



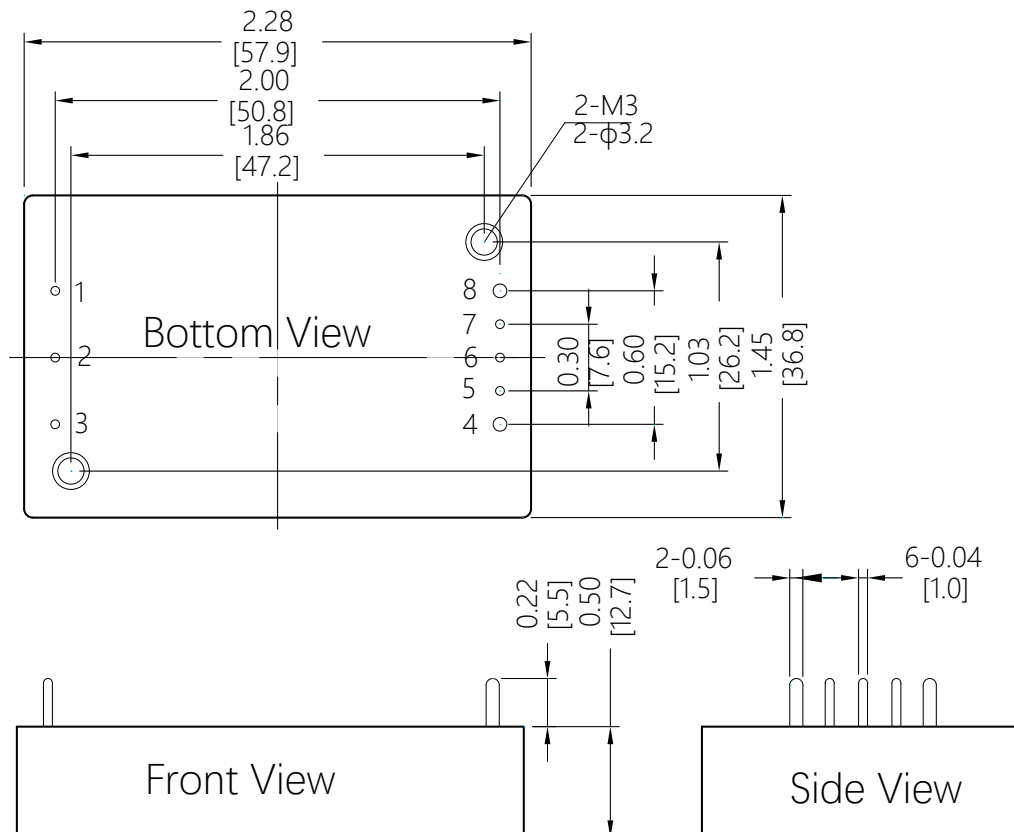
DQB100W110S15 $C_{out}=100\mu F$



DQB100W110S24 $C_{out}=100\mu F$



Mechanical Specifications



PIN:

PIN1, PIN2, PIN3, PIN5 PIN6, PIN7: $\Phi 0.04$ inch

Force: Applied force not exceed 4.9N

PIN4, PIN8 : $\Phi 0.059$ inch

Force: Applied force not exceed 9.8N

Material: Copper alloy

Finish: Gold 3 ~ 5 μ m(min.) over nickel 50 μ m(Min.)

TOLERANCE:

X.XX= ± 0.02 [0.5]

X.XXX= ± 0.010 [0.25]

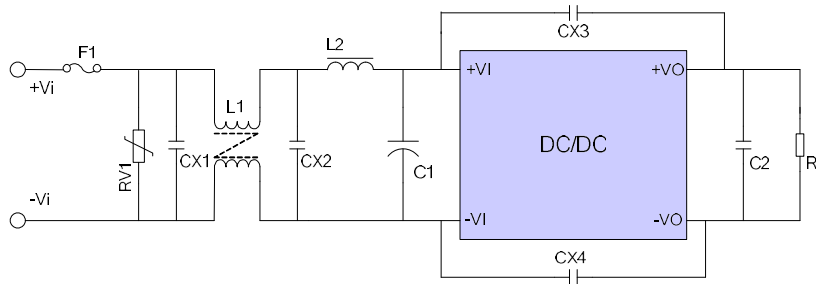
Dimensions are in inches [mm]

Weight: ~66g.

PIN CONNECTIONS	
Single Output	
Pin	Function
1	-Vin
2	RC
3	+Vin
4	+Vout
5	+Sense
6	TRIM
7	-Sense
8	-Vout

Emissions Performance

Densitypower measures its products for conducted emissions against the EN50121-3-2 standards. The common mode filter is added at the output of the module, and the maximum output power of the module is 100W. Input voltage is 110VDC, EMI filter is added outside the modules and the conduction limit can meet class A.

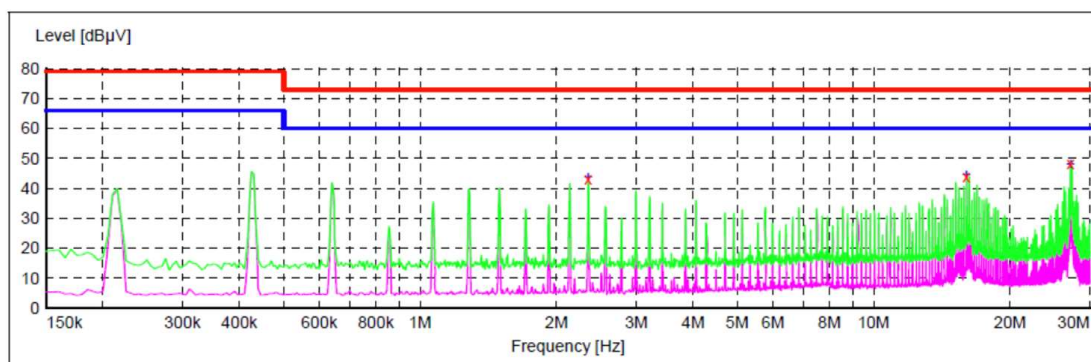


Conducted Emissions Test Circuit

Recommended Filter Parameters

REFERENCE	DESCRIPTION	REFERENCE	DESCRIPTION
F1	15A,250VRetarding fuse	CY3, CY4	222 Y2 CAP
C1, C2, C3	1µF/275VAC	CY5, CY6	102 Y2 CAP
CY1, CY2	102 Y2 CAP	CE1	100uF/250V
L1	1.8mH	CE2, CE3	100µF
L2	2.4mH	L3	70µH

Conducted Emission Test Results: DQB100W110S24



Technical Notes

INPUT FUSING

Certain applications may require fuse at the inputs of power conversion components. Fuses should also be used when there is possibility of sustained input voltage reversal which is not current limited. The DQB100W110 modules are not internally fused. We strongly recommend a slow-blown fuse to be used in the ungrounded input supply line.

For safety agency approvals, the installer must install the converter in compliance with the end user safety standard.

TYPICAL APPLICATION CONNECTION

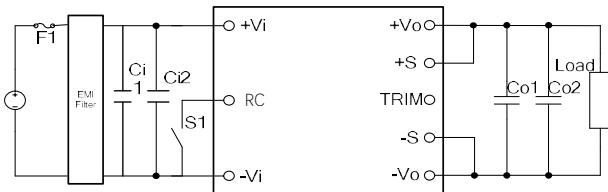


Figure 1: Typical Application Connection

Figure 1 shows the typical use of the module connection. In order to prevent the input line from causing the input oscillation, it is recommended to add the input capacitor close to the input pin of the module. Similarly, the output capacitor is added to the output of the module. Specific recommended parameters: input capacitance $Ci1=22\mu\text{F}$ electrolytic capacitor, $Ci2 = 1\mu\text{F}$ CBB capacitor. Output Capacitance $Co1=10\mu\text{F}$ tantalum capacitor, $Co2 \text{ ESR} < 0.1\Omega$.

REFLECTED RIPPLE CURRENT

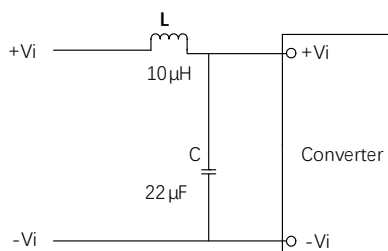


Figure 2: Reflected Ripple Current
Add LC filter at the front of the power module to

reduce the interference of reflected ripple current on the DC bus, recommended value of L and C with appropriate current and voltage rating as below: $L=4.7\mu\text{H}$; $C=100\mu\text{F}$.

REMOTE CONTROL FUNCTION

Module Power Remote Control or called ON/OFF pin is for the user to enable or disable the output. Control use high and low level control, there are two general control logic, positive logic or negative logic control. Recommend to use optocoupler to control ON/OFF Pin as below.

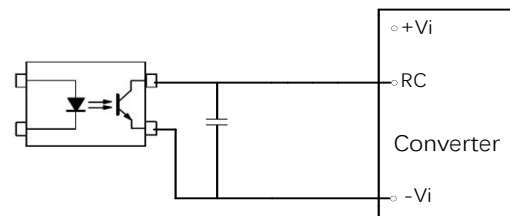


Figure 3: Remote Control

Remote Control Pin can be connected in parallel for multiple converters which with the same Remote Control characters. However, when several converters share the same remote control circuit, the total sink and source current must be taken into consideration, and make sure that the optocoupler has enough drive capability.

To reduce external PCB trace interference, it is recommended to add high frequency bypass capacitor between RC pin and -Vi, recommended capacitor value is 100-1000pF.

REMOTE COMPENSATION FUNCTION

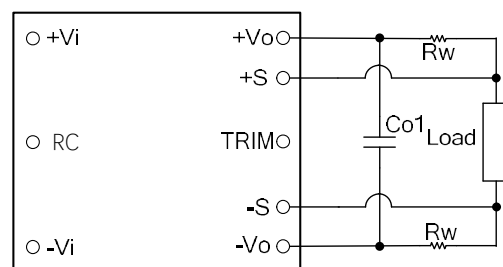


Figure 4: Remote Compensation

Technical Notes

The remote compensation function compensates for the voltage drop across the output line. Module compensation function can't exceed 10%, that is:

$$[(+V_o) - (-V_o)] - [(+S) - (-S)] \leq 10\%V_{onom}$$

If the remote compensation function is not used, the +Sense and +Vout pin, -Sense and -Vout pin need to be shorted directly close to the output.

OUTPUT RIPPLE & NOISE

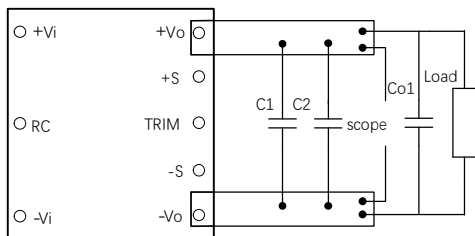


Figure 5: Output Ripple

These DQB100W110 modules' output ripple and noise is measured at the rated input voltage and output current, along with 10uF and 0.1uF MLCC used in parallel with appropriate voltage ratings and placed as C1&C2 shown in the figure above. The scope's bandwidth is set to 20MHz.

External output capacitors are required to reduce the ripple & noise. The output capacitors should be low ESR and appropriate frequency response with appropriate voltage ratings, and must be located as close to the converters as possible, also PCB layout must be taken into consideration.

INPUT UNDERVOLTAGE SHUTDOWN AND START-UP THRESHOLD

Under normal start-up conditions, module will not begin to regulate until the ramping-up input voltage exceeds the Start-Up Threshold Voltage. Once operating, module will not turn off until the input voltage drops below the Undervoltage Shutdown limit. Subsequent re-start will not occur until the input is brought back up to the Start-

Up Threshold. This built in hysteresis prevents any unstable on/off situations from occurring at a single input voltage.

CURRENT LIMITING

The output voltage remains constant as the output current increases. However, once the output current is over the specified Output DC Current Limit, the converter turns off.

The converter then enters a "hiccup mode" where it repeatedly turns on and off until the short circuit condition is removed. This prevents excessive heating of the converter or the load board.

SHORT CIRCUIT CONDITION

When the converter is in current-limit mode, the output voltage will drop as the output current demand increases and then the converter will be shut down. If the short-circuit condition persists, another shutdown cycle will be initiated. This on/off cycling is referred to as "hiccup" mode. The hiccup cycling reduces the average output current, thereby preventing internal temperatures from rising to excessive levels. The module is capable of enduring an indefinite short circuit output condition.

OUTPUT OVERVOLTAGE PROTECTION

The output voltages are monitored for an overvoltage condition via magnetic feedback. The signal is coupled to the primary side and if the output voltage rises to a level which could be damaging to the load, the sensing circuitry will power down the PWM controller causing the output voltages to decrease. Following a timeout period the PWM will restart, causing the output voltages to ramp to their appropriate values. If the fault condition persists, and the output voltages again climb to excessive levels, the overvoltage circuitry will initiate another shutdown cycle. This on/off cycling is referred to as "hiccup" mode.

Technical Notes

THERMAL SHUTDOWN

These DQB100W110 converters are equipped with thermal shutdown circuitry. If environmental conditions cause the internal temperature of the DC-DC converter to rise above the designed operating temperature, a precision temperature sensor will power down the unit. When the internal temperature decreases below the threshold of the temperature sensor, the unit will auto restart.

TRIMMING OUTPUT VOLTAGE

The converters have a trim capability that allow users to adjust the output voltages. Output voltage can be trimmed up or down by a trim pin by connecting a single fixed resistor between Trim Pin and Vout+ or Vout-, the output volatge can be increased or decreased depending on its connection. The maximum output voltage adjustment range is -10% to 10%. If the trim function is not used, keep TRIM pin floating.

Trim up:

Increase the output voltage by connecting an appropriate value resistor between Trim Pin and -Vo Pin. Show as below:

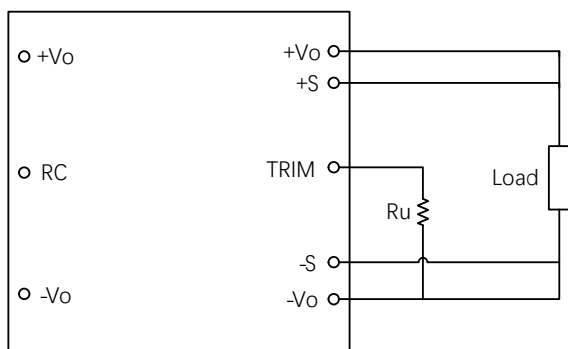


Figure 6· Trim Up Connection

Please follow up the Trim up formula to calculate the resistor value according to the desired output voltage.

$$R_u = R1 \times \frac{V_{ref} - \frac{0.46 \times R2}{R2 + R3}}{\Delta \times V_{oset}} - \frac{R2 \times R3}{R2 + R3}$$

"Voset "is the output voltage when TRIM is floating, Vref is the internal reference voltage value, " Δ%" is the change of output voltage, such as: 15V output is raised to 16.5V.

$$\Delta\% = (16.5-15) / 15 * 100\% = 10\%$$

Trim down:

Decrease the output voltage by connecting an appropriate value resistor between Trim Pin and +Vo Pin. Show as below:

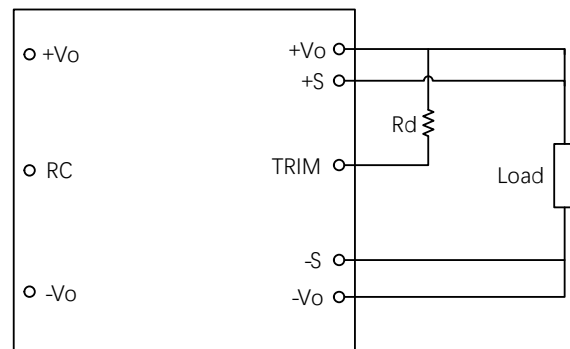


Figure 7· Trim Down Connection

Please follow up the Trim Down formula to calculate the resistor value according to the desired output voltage.

$$R_d = R1 \times \frac{V_{oset} - \Delta \times V_{oset} - V_{ref}}{\Delta \times V_{oset}} - R2$$

"Voset" is the output voltage when TRIM is floating, Vref is the internal reference voltage value, "Δ%" is the amount of change in output voltage, such as: 15V output is reduced to 13.5V, Δ% = (15-13.5) / 15 * 100% = 10%.

R1, R2 is resistance value, please refer to the table below for details.

Technical Notes

Value of R1, R2, R3 and Vref are shown below:

Vout (V)	R1 (K Ω)	R2 (K Ω)	R3 (K Ω)	Vref (V)
5	8.25	5.11	+ ∞	1.24
12	20	100	5.6	1.24
15	11.5	33	5.6	2.5
24	20	100	5.6	2.5
48	36	200	5.1	2.5



This product is subject to the following operating requirements and the Life and Safety Critical Application Sales Policy:

Refer to: <http://www.densitypower.com>

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